

Frequency Modulation (FM):

$$f_i(t) = f_c + \frac{K_f}{2\pi} m(t)$$

$$\theta(t) = 2\pi f_c t + K_f \int_0^t m(\lambda) d\lambda \cdot$$

Where K_f is the deviation sensitivity for frequency or frequency deviation constant for FM

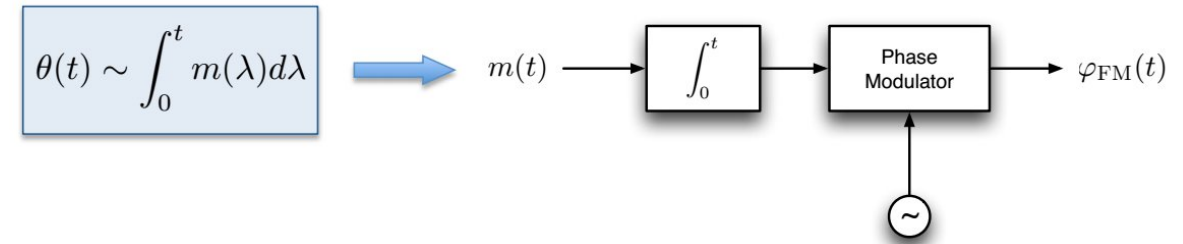
The modulated FM signal is given by

$$\varphi_{\text{FM}}(t) = A_c \cos\left(2\pi f_c t + K_f \int_0^t m(\lambda) d\lambda \right)$$

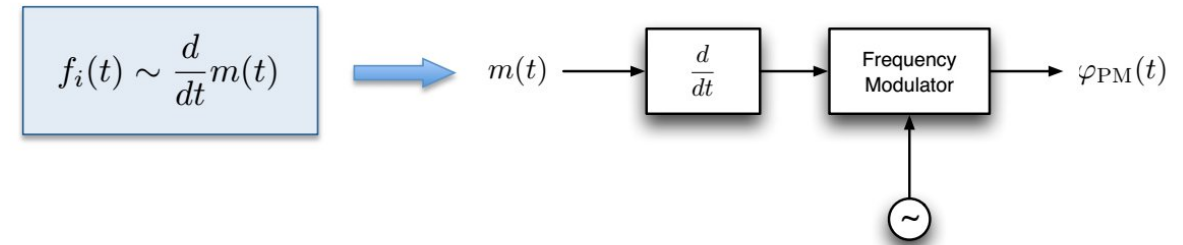
Information contained in $m(t)$ is embedded in the instantaneous frequency:

$f_i(t)$ is proportional to $m(t)$

PM and FM are tightly related to each other. We see from the phase and frequency relations for PM and FM given above that replacing $m(t)$ in the PM signal with $\int m(\lambda) d\lambda$ gives an FM signal and replacing $m(t)$ in the FM signal with $\frac{d m(t)}{dt}$ gives a PM signal. This is illustrated in the following block diagrams.



or we can consider PM as a special case of FM where:



Example : An angle modulated signal is given by $x(t) = \cos(2000t + 0.1\sin(20t))$

- Find the frequency of the modulated carrier f_c
- Instantaneous frequency $f_i(t)$

Solution

a) $2\pi f_c = 2000, f_c = 318.31\text{Hz}$.

b) $\varphi(t) = 2000t + 0.1\sin(20t)$

$$f_i(t) = \frac{1}{2\pi} \frac{d\theta(t)}{dt} = \frac{1}{2\pi} (2000 + 2 \cos(20t)) = 318.31 + 0.32 \cos(20t)$$

Assume $m(t)$ is information signal.

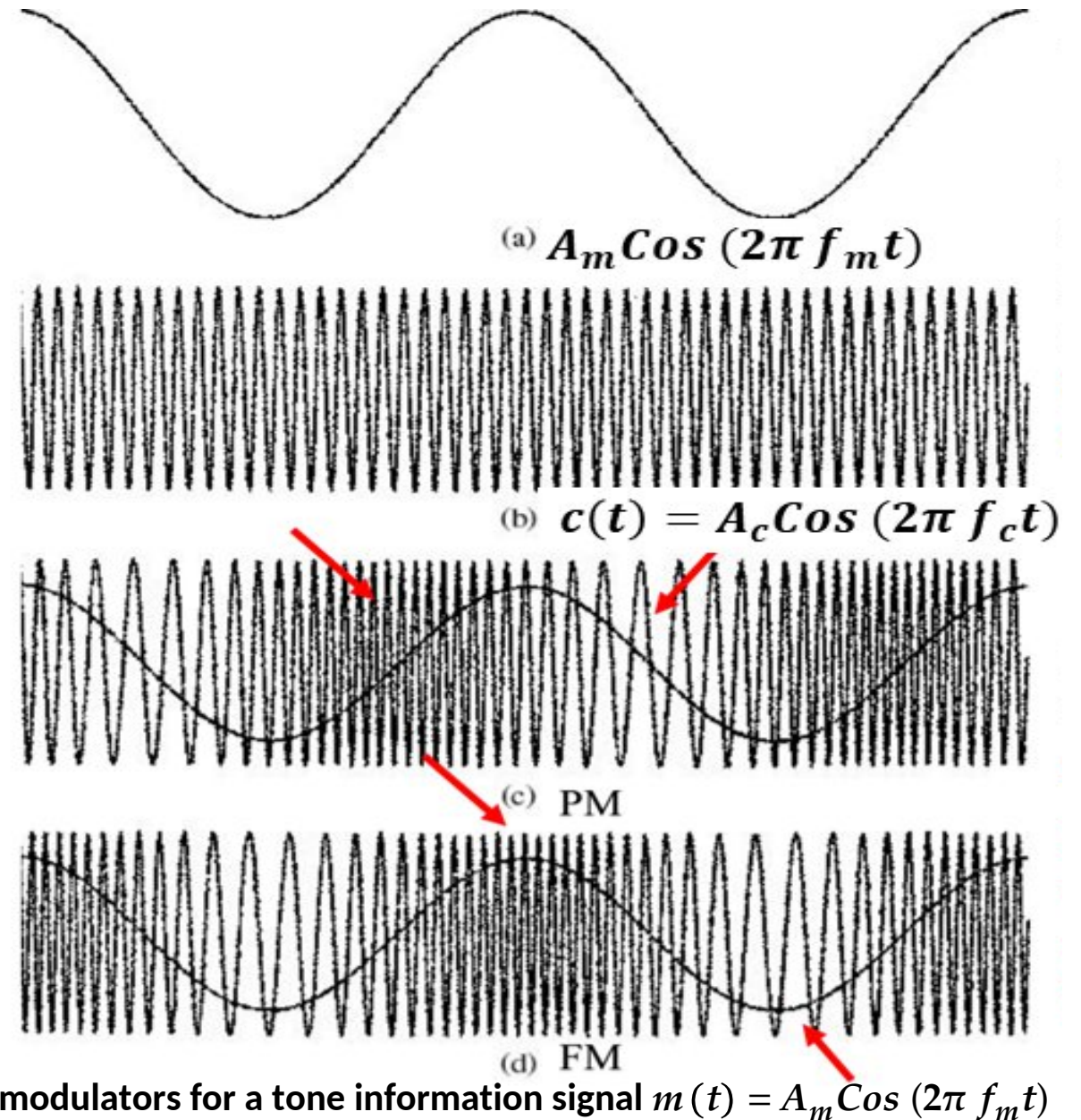
For PM

The phase deviation of modulator output is proportional to $m(t)$. The frequency deviation is proportional to the **derivative** of the phase deviation, thus the ω_i is maximum when the **slope** of $m(t)$ is maximum and minimum when the **slope** of $m(t)$ is minimum.

For FM

Frequency deviation is proportional to $m(t)$, thus the ω_i is max when the $m(t)$ is maximum and it is minimum when $m(t)$ is minimum.

Note that if $m(t)$ is not shown along with the modulator o/p. It would not be possible to distinguish the PM and FM o/p



Example : A phase modulated signal has a carrier of frequency 2kHz. Draw the phase modulated signal given the phase sensitivity is 2 volt⁻¹ for the below modulating signal m(t).

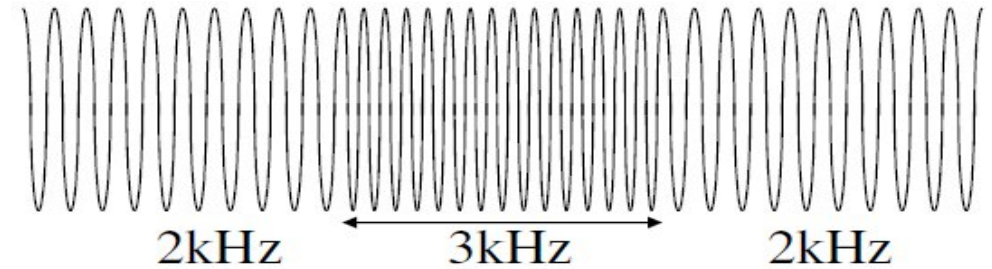
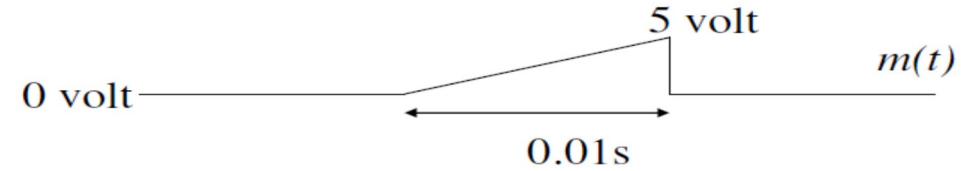
Solution

$$\varphi_{PM}(t) = A_c \cos(2\pi f_c t + K_p m(t))$$

$$\theta(t) = 2\pi f_c t + K_p m(t)$$

The instantaneous frequency is

- $f_i = f_c + 2(0) = 2\text{kHz}$, for $m(t) = 0$.
- $f_i = f_c + 2(500) = 3\text{kHz}$, for $m(t) = 500t$.



Example : A phase modulated signal has a carrier of 100 MHz and the modulating signal is 1 kHz sinusoid of amplitude 0.01 V. The proportionality constant, K_p , is 10^4 V⁻¹. Find the band of frequencies occupied by the PM waveform.

Solution: PM

$$\begin{aligned} \theta(t) &= 2\pi f_c t + k_p m(t) \\ &= 2\pi 10^8 t + 10^4 \times 0.01 \cos(2\pi \times 10^3 t) \end{aligned}$$

- The instantaneous frequencies,

$$f_i(t) = \frac{1}{2\pi} \frac{d\theta}{dt} = 10^8 - 2\pi \times 10^5 \sin(2\pi \times 10^3 t)$$

Example : Tone signal $m(t) = A_m \cos(\omega_m t)$ is given. Find FM and PM waveforms.

Solution $\varphi(t) = A_c \cos \theta(t)$

$A_c = \text{constant}$: we need to find $\theta(t)$

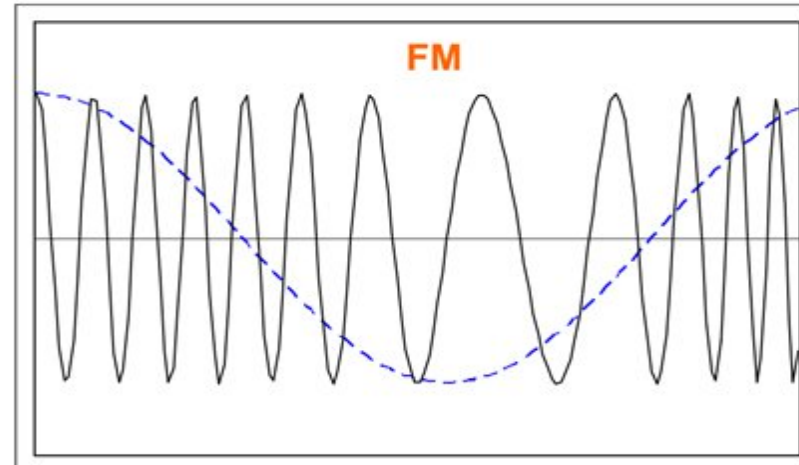
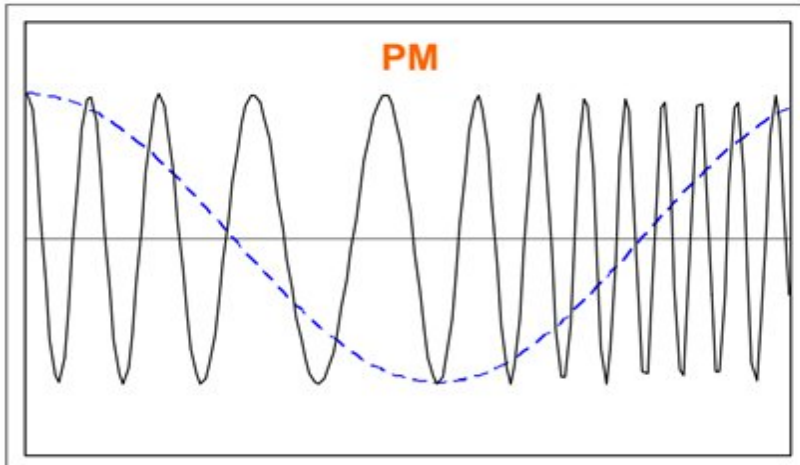
$$\theta(t) = 2\pi f_c t + K_p m(t) + \theta_0 = \omega_c t + k_p m(t) = \omega_c t + k_p A_m \cos(\omega_m t) \quad (\text{PM})$$

$$\theta(t) = 2\pi f_c t + k_f \int_{-\infty}^t m(\tau) d\tau = 2\pi f_c t + k_f \int_{-\infty}^t A_m \cos(\omega_m \tau) d\tau = \omega_c t + \frac{k_f A_m}{\omega_m} \sin(\omega_m t) \quad (\text{FM})$$

$$\varphi_{\text{PM}}(t) = A_c \cos(\omega_c t + k_p A_m \cos(\omega_m t)) \quad (\text{PM})$$

therefore

$$\varphi_{\text{FM}}(t) = A_c \cos(\omega_c t + \frac{k_f A_m}{\omega_m} \sin(\omega_m t)) \quad (\text{FM})$$



One primary difference between PM and FM is the way the modulation index μ is defined

- **For PM**, μ is proportional to the amplitude of the modulating signal (A_m) and independent of its frequency

$$\mu_p = K_p A_m \quad (\text{unit in radians}) \quad (\text{equal to Peak phase deviation } \Delta\theta)$$

- **For FM**, μ is proportional to the amplitude of the modulating signal (A_m) and inversely proportional its frequency (f_m)

$$\mu_f = \frac{K_f A_m}{f_m} \quad (\text{unitless})$$

Thus for tone signal the modulated signal can be rewritten as

$$\varphi_{\text{PM}}(t) = A_c \cos(\omega_c t + \mu_p \cos(\omega_m t)) \quad (\text{PM})$$

$$\varphi_{\text{FM}}(t) = A_c \cos(\omega_c t + \mu_f \sin(\omega_m t)) \quad (\text{FM})$$

Frequency Deviation – the change in frequency that occurs in the carrier when it is acted on by a modulating-signal frequency.

- Typically given as a peak frequency shift in hertz (Δf).
- The peak-to-peak frequency deviation ($2 \Delta f$) is called carrier swing.
- For FM, the deviation sensitivity, K_f is often given in hertz/volt.

Thus, the peak frequency deviation can be expressed mathematically as:

$$\Delta f = K_f A_m \text{ (Hz)}$$

Therefore the modulation index can be expressed as

$$\mu_f = \frac{\Delta f}{f_m} \text{ (unitless)}$$

Percent Modulation – the ratio of actual frequency deviation to the maximum frequency deviation allowed in percentage form.

$$\% \text{ modulation} = \frac{\Delta f \text{ (actual)}}{\Delta f \text{ (Max)}} \times 100$$

Example Determine the peak frequency deviation (Δf) and modulation index μ for an FM modulator with a deviation sensitivity, $K_f = 10\text{kHz/V}$ and a modulating signal $m(t) = 5 \cos(2\pi \times 2500t)$.

Solution

1. $\Delta f = K_f A_m = \frac{10 \text{ KHz}}{\text{V}} (5\text{V}) = 50 \text{ KHz}$
2. $\mu_f = \frac{\Delta f}{f_m} = \frac{50\text{KHz}}{2500} = 20$

Example Determine the peak phase deviation ($\Delta\theta$) for a PM modulator with a deviation sensitivity $K_p = 2.5\text{rad/V}$ and a modulating signal $m(t) = 2 \cos(2\pi(2000)t)$.

Solution

$$\Delta\theta = K_p A_m = 2.5 \left(\frac{\text{rad}}{\text{V}}\right) \times 2\text{V} = 5 \text{ radians}$$

Example The carrier frequency in an FM modulator is 1000 kHz. If the modulating frequency is 15 kHz, what are the first three upper sideband and lower sideband frequencies?

Solution.

Carrier frequency, $f_c = 1000$ kHz

Modulating frequency, $f_m = 15$ kHz

Upper sideband frequencies

$$\begin{array}{lll} f_c + f_m & ; & f_c + 2f_m & ; & f_c + 3f_m \\ 1000 + 15 & ; & 1000 + 2 \times 15 & ; & 1000 + 3 \times 15 \\ \mathbf{1015 \text{ kHz}} & ; & \mathbf{1030 \text{ kHz}} & ; & \mathbf{1045 \text{ kHz}} \end{array}$$

Lower sideband frequencies

$$\begin{array}{lll} f_c - f_m & ; & f_c - 2f_m & ; & f_c - 3f_m \\ 1000 - 15 & ; & 1000 - 2 \times 15 & ; & 1000 - 3 \times 15 \\ \mathbf{985 \text{ kHz}} & ; & \mathbf{970 \text{ kHz}} & ; & \mathbf{955 \text{ kHz}} \end{array}$$

- Theoretically, the generation and transmission of FM requires infinite bandwidth. Practically, FM system have finite bandwidth and they perform well.
- The value of **modulation index** determine the number of sidebands that have the significant relative amplitudes
- If n is the number of sideband pairs, and line of frequency spectrum are spaced by f_m , thus, the bandwidth is:

$$B_{FM} = 2n f_m \quad \text{For } n \geq 1$$

Carson's Rule for FM Bandwidth.

An approximation for the bandwidth of an FM signal is given by $BW = 2(\text{Maximum frequency deviation} + \text{highest modulating frequency})$

$$\begin{aligned} B_{FM} &= 2(m_f + 1) f_m \\ &= 2(Df + f_m) \end{aligned}$$

The same rule applies to PM bandwidth,

$$B_{PM} = 2(Df + f_m)$$

For PM the $\Delta f = \frac{1}{2\pi} K_p \left| \frac{d m(t)}{dt} \right|_{\max}$

$$\Delta f = \Delta \omega / 2\pi$$

When $\Delta f \ll f_m$, modulation index < 1 , the scheme is called Narrow Band (NBFM, NBPM).

Then the bandwidth is

$$B_{NBFM} = B_{NBPM} = 2(f_m)$$

Therefore, no matter how small we make the deviation around f_c , the bandwidth of the modulated signal does not get smaller than $2f_m$ or $2B_m$.

Example: If the modulating frequency is 1 kHz and the maximum deviation is 10 kHz, what is the modulation index and the required bandwidth for an FM signal?

Solution

$$f_m = 1\text{kHz}, \Delta f = 10\text{kHz}$$

$$\mu_f = \frac{\Delta f}{f_m} = \frac{10\text{kHz}}{1\text{kHz}} = 10$$

$$\begin{aligned} B_{FM} &= 2(m_f + 1) f_m \\ &= 22\text{kHz} \end{aligned}$$

Example : Consider the angle modulated (PM) signal $\phi(t) = 10 \cos(2\pi 10^8 t + 3 \sin 2\pi 10^3 t)$. Find the bandwidth of the PM signal when the message frequency is (1) doubled, and (2) is halved.

Solution

$$\begin{aligned} \phi_{PM}(t) &= A_c \cos(\theta(t)) = 10 \cos(2\pi 10^8 t + 3 \sin 2\pi 10^3 t) \\ &= A_c \cos(2\pi f_c t + K_p m(t)) \end{aligned}$$

We have $f_m = 1 \text{ kHz}$ and $K_p = 3$

$$\Delta\omega \text{ for PM} = K_p \left| \frac{d m(t)}{dt} \right|_{\max} = 6\pi \text{ kHz} \quad : \Delta f = \Delta\omega / 2\pi = 3 \text{ kHz}$$

$$B_{PM} = 2(\Delta f + f_m) = 2(3 + 1) = 8 \text{ kHz}$$

When the message frequency is doubled

$$\Delta\omega \text{ for PM} = K_p \left| \frac{d m(t)}{dt} \right|_{\max} = 12 \pi \text{ kHz}$$

$$B_{PM} = 2(\Delta f + f_m) = 2(6 + 2) = 16 \text{ kHz}$$

When the message frequency is halved $\Delta f = 1.5 \text{ kHz}$

$$B_{PM} = 2(\Delta f + f_m) = 2(1.5 + 0.5) = 4 \text{ kHz}$$

$$m(t) = \sin 2\pi 10^3 t$$

$$\frac{d m(t)}{dt} = 2\pi 10^3 \cos(2\pi 10^3 t)$$

$$\left| \frac{d m(t)}{dt} \right|_{\max} = 2\pi 10^3$$

Deviation Ratio (DR) is defined as the worst-case modulation index and is equal to the maximum peak frequency deviation divided by the maximum modulating-signal frequency.

$$DR = \frac{\Delta f_{max}}{f_{m(max)}}$$

The worst-case modulation index produces the widest output frequency spectrum.

Example : Determine the deviation ratio and the bandwidth for an FM broadcast-band transmitter with a maximum frequency deviation of 75kHz and a maximum modulating-signal frequency of 15kHz.

Solution:

$$DR = \frac{\Delta f_{max}}{f_{m(max)}} = \frac{75\text{KHz}}{15\text{KHz}} = 5$$

$$B_{FM} = 2(\Delta f + f_m) = 2(75+15) = 180\text{KHz}$$

Some mobile phone companies use FM with a very low modulation index, i.e. $\mu_f < 1$. This is known as **narrowband FM**. Mobile phone companies use this because it offers many of the advantages of FM, with the minimum bandwidth requirement.

Advantages

- Wideband FM gives significant *improvement in the SNR* at the output of the RX which is proportional to the square of modulation index.
- Angle modulation is *resistant to propagation-induced* selective fading since amplitude variations are unimportant and are removed at the receiver using a limiting circuit.
- Angle modulation is very *effective in rejecting interference*. (minimizes the effect of noise).
- Angle modulation allows the use of *more efficient* transmitter power in information.
- Angle modulation is capable of *handling a greater dynamic range* of modulating signal without distortion than AM.

Disadvantages

- Angle modulation requires a transmission bandwidth *much larger* than the message signal bandwidth.
- Angle modulation requires *more complex* and expensive circuits than AM.

Questions?

**Thank you for
Listening & Attending**